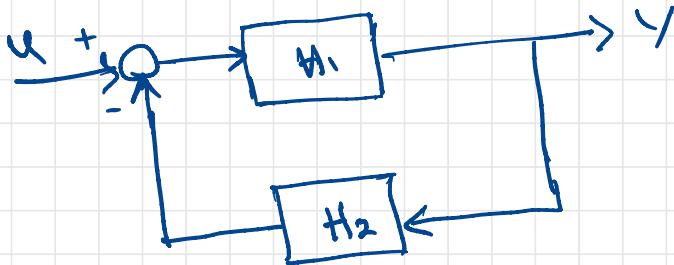


Absolute stability:



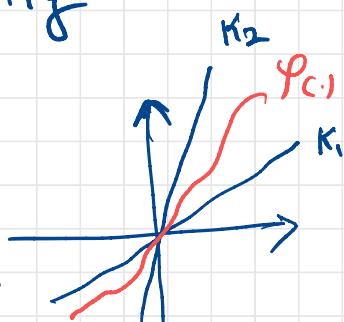
- Passivity theorem:

if H_1 and H_2 are passive, then the FB sys. is
passive \Rightarrow stable closed loop

- We study the special case where H_1 is linear
and H_2 is a sector nonlinearity

$$H_1 : \begin{aligned} \dot{x} &= Ax + Bu \\ y &= Cx + Du \end{aligned}$$

$$H_2 : y = \varphi(u), \varphi \in [k_1, k_2]$$



- Problem: find conditions for H_1 s.t. the
FB sys. is stable.

- We use passivity to answer this question.

Passivity for linear systems:

$$\begin{aligned}\dot{x} &= Ax + Bu \\ y &= Cx + Du\end{aligned} \iff \begin{aligned}\dot{y}(s) &= G(s)\vec{u}(s)\end{aligned}$$

↓
transfer function

- $x \in \mathbb{R}^n$, $y, u \in \mathbb{R}^p$

- To discuss passivity for lin. sys. we need to introduce the notion of positive real transfer functions.

Def:

- When $p=1$ (SISO), $G(s)$ is positive real if
 - 1) all poles are in LHP ($\operatorname{Re}(s) \leq 0$)
 - 2) $\operatorname{Re}(G(j\omega)) \geq 0 \quad \forall \omega$ s.t. $j\omega$ is not a pole
 - 3) if $j\omega$ is a pole, then it is a simple pole and $\lim_{s \rightarrow j\omega} (s - j\omega) G(s) \geq 0$

- $G(s)$ is strictly positive real if $G(s-\varepsilon)$ is positive real for some $\varepsilon > 0$

- The definition can be extended to MIMO
(See Def 6.4)

Examples:

a) $G(s) = \frac{1}{s}$ \rightsquigarrow only pole $s=0$

\Rightarrow 1) all poles in LHP \checkmark

2) $\operatorname{Re}(G(j\omega)) = \operatorname{Re}\left(\frac{1}{j\omega}\right) = 0 \quad \checkmark$

3) $\lim_{s \rightarrow 0} sG(s) = 1 \geq 0 \quad \checkmark$

\Rightarrow positive real

but not strictly positive real because

$$G(s-\varepsilon) = \frac{1}{s-\varepsilon} \text{ has pole } s=\varepsilon > 0$$

$$b) G(s) = \frac{1}{s+a} \text{ for } a > 0$$

this is positive real

1) pole $s = -a$ in LHS ✓

$$2) \operatorname{Re}(G(j\omega)) = \operatorname{Re}\left(\frac{1}{j\omega+a}\right) = \frac{a}{a^2+\omega^2} \geq 0$$

No imaginary pole

Also, strictly positive real because

for $\epsilon = a > 0$, $G(s-a) = \frac{1}{s}$ is positive real.

$$c) G(s) = \frac{1}{s^2+s+1}$$

when $|w| > 1$

$$G(j\omega) = \frac{1}{j\omega + 1 - \omega^2} \Rightarrow \operatorname{Re}(G(j\omega)) = \frac{1-\omega^2}{(1-\omega^2)^2 + \omega^2} < 0$$

⇒ not positive real.

- In general, if $G(s) = \frac{P(s)}{Q(s)}$ is positive real

then relative degree = $\deg(Q) - \deg(P) \leq 0$ or 1

Positive real lemma: (Lemma 6.2, 6.3)

- Assume (A, B) is controllable and (A, C) is observable
- Transfer function $G(s) = C[sI - A]^{-1}B + D$ is positive real iff \exists a p.d matrix P and matrices L, W , s.t.

$$PA + A^T P = -L^T L$$

$$PB = C^T - L^T W \iff$$

$$W^T W = D + D^T$$

$$\begin{bmatrix} PA + A^T P, & PB - C^T \\ B^T P - C, & D + D^T \end{bmatrix} \leq 0$$
$$= - \begin{bmatrix} L^T \\ W^T \end{bmatrix} [L, W]$$

- Strictly positive real if for some $\varepsilon > 0$

$$PA + A^T P = -L^T L - \varepsilon P$$

$$PB = C^T - L^T W$$

$$W^T W = D + D^T$$

Lemma 5.4:

(Strict) positive real \Leftrightarrow (strict) passivity

Proof:

- Consider $V = \frac{1}{2} X^T P X$, where P is from positive real lemma.

$$\dot{V} = \frac{1}{2} X^T (PA + A^T P) X + X^T P B u$$

$$PA + A^T P = -L^T L$$

$$PB = C^T - L^T W$$

$$= -\frac{1}{2} X^T L^T L X + X^T C^T u - X^T L^T W u$$

$$= -\frac{1}{2} (LX + Wu)^T (LX + Wu)$$

$$+ \frac{1}{2} U^T W^T W u + X^T C^T u$$

$$W^T W = D + D^T$$

$$C X = Y - D u$$

$$\leq \frac{1}{2} U^T (D + D^T) U + Y^T u - U^T D^T u$$

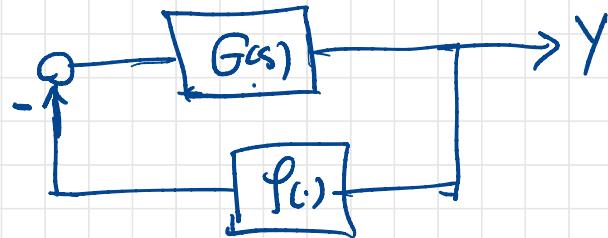
$$= Y^T u \quad \Rightarrow \text{passive}$$

- The proof for strict passivity is similar

$$\dot{V} \leq Y^T u - \frac{\varepsilon}{2} X^T P X$$

Back to absolute-stability: (circle criteria)

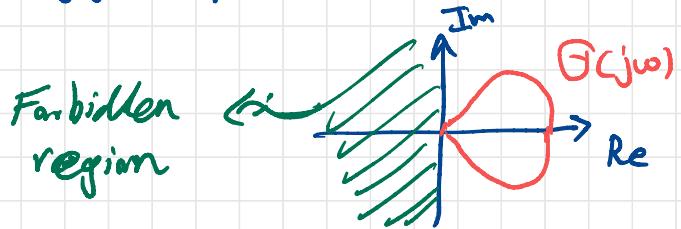
- Assume $\varphi \in [0, \infty]$



- Therefore $P(s)$ is passive

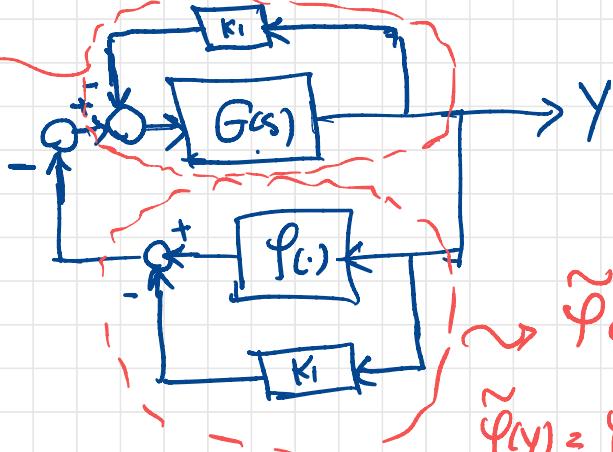
- The FB sys. is stable if $G(s)$ is passive.

- Pictorially, the Nyquist plot is in the right-hand plane.



- Now, if $\varphi \in [K_1, \infty]$, we do a loop transformation to turn φ to $\tilde{\varphi} \in [0, \infty]$

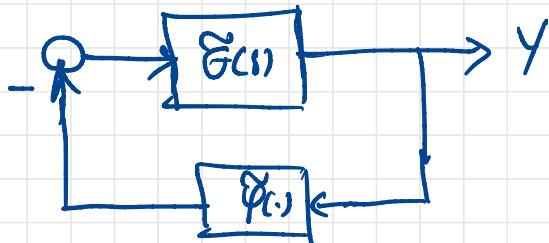
$$\tilde{G}(s) = \frac{G(s)}{1 + K_1 G(s)}$$



$$\tilde{\varphi}(y) = \varphi(y) - K_1 y$$

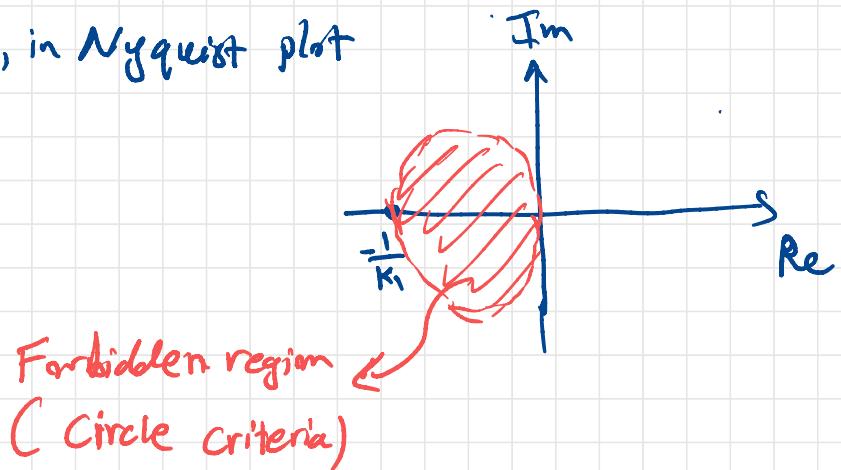
$$\begin{aligned} \tilde{\varphi} &\in [0, \infty] \\ \tilde{\varphi}(y) &= \varphi(y) - K_1 y \end{aligned}$$

- Equivalent loop



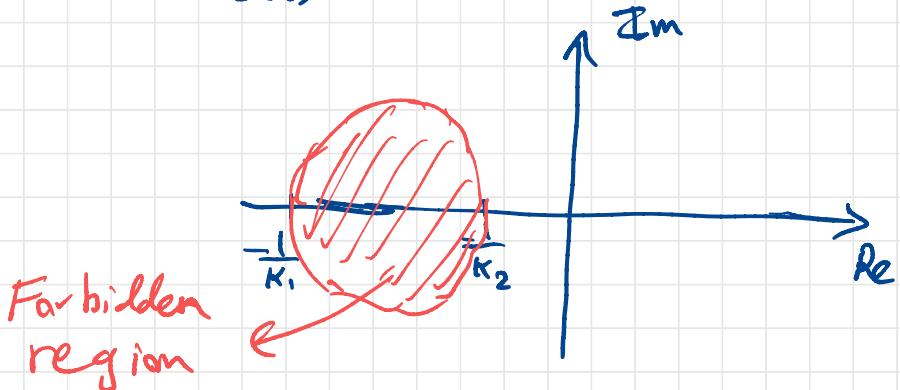
- So, to have a passive FB sys. $\tilde{G}(s) \equiv \frac{G(s)}{1+K_1 G(s)}$ should be positive real

- Pictorially, in Nyquist plot



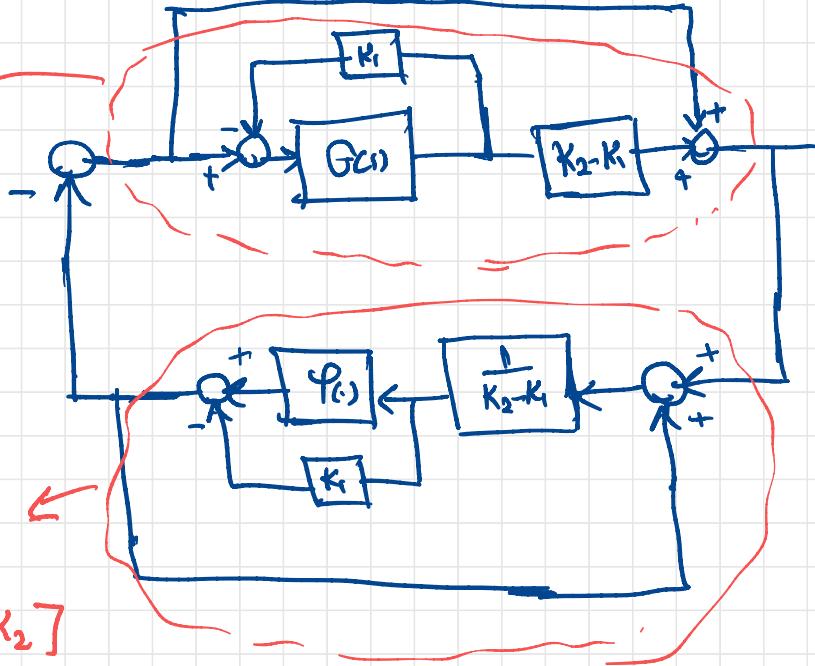
- More generally, if $\varphi \in [K_1, K_2]$, then with loop transformation, we have

$$\tilde{G}(s) = \frac{1 + K_2 G(s)}{1 + K_1 G(s)} \text{ to be positive real.}$$



Loop trans. ?

$$\tilde{G}(s) = \frac{1 + K_2 G(s)}{1 + K_1 G(s)}$$



$$\varphi \in [0, \infty]$$

when $\varphi \in [K_1, K_2]$